

EE 233 HW#6 Solution

(13.70) (a)

$$\begin{aligned}
 H(s) &= \frac{\frac{1}{sC}}{\frac{1}{sC} + R + sC} \\
 &= \frac{1}{s^2LC + sCR + 1} \\
 &= \frac{\frac{1}{LC}}{s^2 + s\frac{R}{L} + \frac{1}{LC}} \\
 &= \frac{100}{s^2 + s20 + 100} \\
 &= \frac{100}{(s + 10)^2} \\
 \mathcal{L}^{-1} \left\{ \frac{100}{(s + 10)^2} \right\} &= te^{-10t} 100 \\
 &= 100te^{-10t}
 \end{aligned}$$

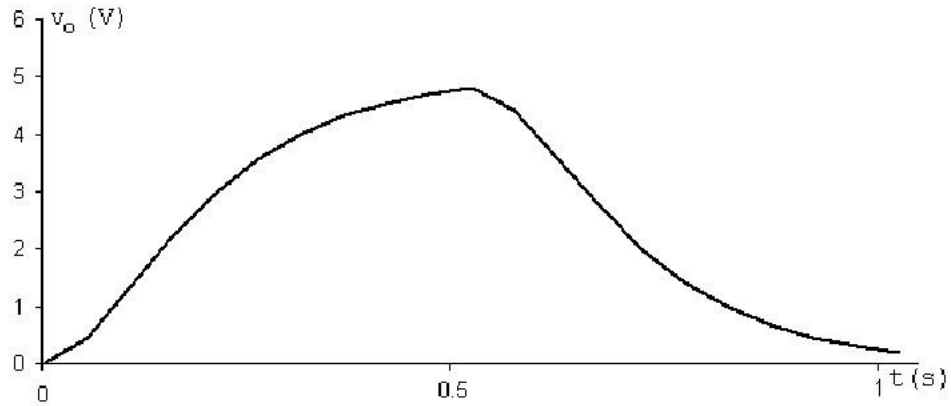
if $V_i = 5[u(t) - u(t - 0.5)]$ So
for $0 \leq t \leq 0.5$:

$$\begin{aligned}
 y(t) &= \int_0^t 5(100)\lambda e^{-10\lambda} d\lambda \\
 &= 5e^{-10\lambda} (10\lambda + 1) \Big|_{\lambda=0}^{\lambda=t} \\
 &= -5e^{-10t} (10t + 1) - (-5e^{-0} (10(0) + 1)) \\
 &= -5e^{-10t} (10t + 1) + 5
 \end{aligned}$$

And for $0.5 \leq t \leq \infty$:

$$\begin{aligned}
 V_0(t) &= \int_{t-0.5}^t 5(100)\lambda e^{-10\lambda} d\lambda \\
 &= -5e^{-10\lambda} (10\lambda + 1) \Big|_{\lambda=t-0.5}^{\lambda=t} \\
 &= -5e^{-10t} (10t + 1) - (-5e^{-10(t-0.5)} (10(t-0.5) + 1)) \\
 &= -5e^{-10t} (10t + 1 - e^5 (10t - 4))
 \end{aligned}$$

(b) The graph is shown here:



(13.80) input of 30, $u(t)$ is applied, the response is

$V_{out} = (50e^{-8000t} - 20e^{-5000t})u(t)$ when $V_{in} = 30u(t)$.

$$\begin{aligned}
 V_o(s) &= \frac{50}{s+8000} - \frac{20}{s+5000} \\
 &= \frac{50(s+5000) - 20(s+8000)}{(s+5000)(s+8000)} \\
 &= \frac{50s + 250000 - 20s - 160000}{(s+5000)(s+8000)} \\
 &= \frac{30s + 90000}{(s+5000)(s+8000)} \\
 &= \frac{30(s+3000)}{(s+5000)(s+8000)}
 \end{aligned}$$

So to find $H(s)$ we need to divide by $\mathcal{L}\{30u(t)\} = \frac{30}{s}$.

$$\begin{aligned}
 H(s) &= \frac{30(s+3000)}{\frac{30}{s}(s+5000)(s+8000)} \\
 &= \frac{s(s+3000)}{(s+5000)(s+8000)}
 \end{aligned}$$

$$V_{in} = 120 \cos(6000t)$$

We know that $s = j\omega$ and $\omega = 6000$, so

$$\begin{aligned}
 H(j6000) &= \frac{(j6000)(3000 + j6000)}{(5000 + j6000)(8000 + j6000)} = 0.206 + 0.472j \\
 &= 0.52 \angle 66.37^\circ
 \end{aligned}$$

Thus, $y(t) = (120)(0.52 \cos(6000t + 66.37))$

$$y(t) = 62.4 \cos(6000t + 66.37)$$

(13.91) We can transform the current source into a voltage source. So the source voltage becomes $(50K)(0.5\delta(t))$ and the resistor in series with the source.

$$C_{eq} = \frac{(1\mu F)(250nF)}{1\mu F + 250nF} = 0.2\mu F$$

$$\begin{aligned}
H(s) &= \frac{0.5 \frac{1}{s(1\mu F)}}{\frac{1}{s(1\mu F)} + \frac{1}{s(250nF)} + 50K} \\
&= \frac{0.5}{1 + \frac{1\mu F}{250nF} + 50000(1\mu F)s} \\
&= \frac{0.5}{5 + .05s} \\
&= \frac{10}{100 + s} \\
\mathcal{L}^{-1} \left\{ \frac{10}{100 + s} \right\} &= 10e^{-100t}u(t) \\
&= V_0(t)
\end{aligned}$$

(b) We would expect the circuit to produce a decaying exponential with $\tau = RC$

$$\frac{1}{(50K\Omega)(0.2\mu F)} = 100$$

So yes, this makes sense.

(Problem1) (a)

$$\begin{aligned}
H(s) &= \frac{\frac{1}{sC}}{\frac{1}{sC} + R} \\
&= \frac{1}{sCR + 1} \\
&= \frac{\frac{1}{RC}}{s + \frac{1}{RC}} \\
&= \frac{250}{s + 250} \\
\mathcal{L}^{-1} \left\{ \frac{250}{s + 250} \right\} &= 250e^{-250t}
\end{aligned}$$

So for $0 \leq t \leq 4$ ms,

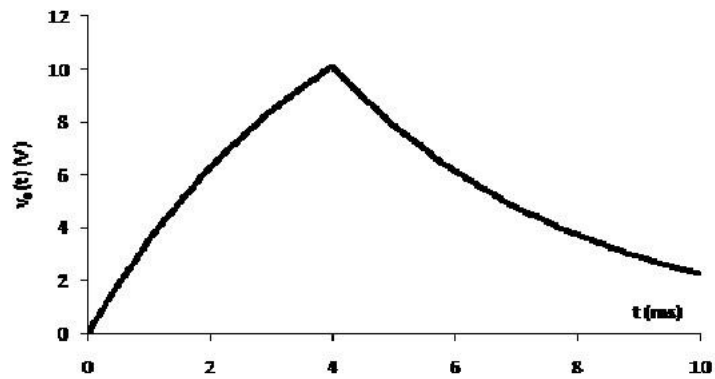
$$\begin{aligned}
V_0(t) &= \int_0^t 16(250)e^{-250\lambda} d\lambda \\
&= -16e^{-250\lambda} \Big|_{\lambda=0}^{\lambda=t} \\
&= 16(1 - e^{-250t})
\end{aligned}$$

and for $4\text{ms} \leq t \leq \infty$:

$$V_0(t) = \int_{t-0.004}^t 16(250)e^{-250\lambda} d\lambda$$

$$\begin{aligned}
 &= -16e^{-250\lambda} \Big|_{\lambda=t-0.004}^{\lambda=t} \\
 &= -16e^{-250t} + 16e^{-250(t-0.004)} \\
 &= 16e^{-250t}(e - 1)V
 \end{aligned}$$

(b) The graph is shown here:



(Problem2)

$$\text{P 14.3 [a]} \quad H(s) = \frac{V_o}{V_i} = \frac{R}{sL + R + R_l} = \frac{(R/L)}{s + (R + R_l)/L}$$

$$\text{[b]} \quad H(j\omega) = \frac{(R/L)}{\left(\frac{R+R_l}{L}\right) + j\omega}$$

$$|H(j\omega)| = \frac{(R/L)}{\sqrt{\left(\frac{R+R_l}{L}\right)^2 + \omega^2}}$$

$$|H(j\omega)|_{\max} \text{ occurs when } \omega = 0$$

because as ω increase, then the denominator will get bigger and decrease the gain

$$\text{[c]} \quad |H(j\omega)|_{\max} = \frac{R}{R + R_l}$$

$$\text{[d]} \quad \frac{|H(j\omega)|_{\max}}{\sqrt{2}} = |H(j\omega_c)|$$

$$|H(j\omega_c)| = \frac{R}{\sqrt{2}(R + R_l)} = \frac{R/L}{\sqrt{\left(\frac{R+R_l}{L}\right)^2 + \omega_c^2}}$$

$$\therefore \omega_c^2 = \left(\frac{R + R_l}{L}\right)^2; \quad \therefore \omega_c = (R + R_l)/L$$

$$\text{[e]} \quad \omega_c = \frac{127 + 75}{0.01} = 20,200 \text{ rad/s}$$

$$H(j\omega) = \frac{12,700}{20,200 + j\omega}$$

$$H(j0) = 0.6287$$

$$H(j20,200) = \frac{0.6287}{\sqrt{2}} \angle -45^\circ = 0.4446 \angle -45^\circ$$

$$H(j6060) = \frac{12,700}{20,200 + j6060} = 0.6022 \angle -16.70^\circ$$

$$H(j60,600) = \frac{12,700}{20,200 + j60,600} = 0.1988 \angle -71.57^\circ$$

(Problem3)

- a) We know center frequency

$$\omega_0 = \sqrt{\frac{1}{LC}} = \frac{20krad}{s}$$

So we can get $L = 0.05H$

Use the definition of quality factor

$$Q = \omega_0 RC = 5$$

We get $R = 5K\Omega$

- b) Lower cutoff frequency:

$$\omega_{c1} = \omega_0 \left[-\frac{1}{2Q} + \sqrt{1 + \left(\frac{1}{2Q}\right)^2} \right] = 18010rad/s$$
$$f_{c1} = \frac{\omega_{c1}}{2\pi} = 2866.4Hz = 2.866KHz$$

Upper cutoff frequency:

$$\omega_{c2} = \omega_0 \left[\frac{1}{2Q} + \sqrt{1 + \left(\frac{1}{2Q}\right)^2} \right] = 22099.8rad/s$$
$$f_{c2} = \frac{\omega_{c2}}{2\pi} = 3517.3Hz = 3.517KHz$$

- c) The bandwidth in this filter is

$$B = f_{c2} - f_{c1} = 3.517KHz - 2.866KHz = 0.651KHz = 651Hz$$